

Audio Engineering Society Convention Paper 8501

Presented at the 131st Convention 2011 October 20–23 New York, NY, USA

This Convention paper was selected based on a submitted abstract and 750-word precis that have been peer reviewed by at least two qualified anonymous reviewers. The complete manuscript was not peer reviewed. This convention paper has been reproduced from the author's advance manuscript without editing, corrections, or consideration by the Review Board. The AES takes no responsibility for the contents. Additional papers may be obtained by sending request and remittance to Audio Engineering Society, 60 East 42nd Street, New York, New York 10165-2520, USA; also see www.aes.org. All rights reserved. Reproduction of this paper, or any portion thereof, is not permitted without direct permission from the Journal of the Audio Engineering Society.

Practical applications of a Closed Feedback Loop Transducer system equipped with Differential Pressure Control

Fabio Blasizzo¹, Paolo Desii², Mario Di Cola³, Claudio Lastrucci²

¹ Fabio Blasizzo, Trieste, Italy <u>ciubexx@inwind.it</u>

² Powersoft S.r.I, Scandicci, Firenze, Italy paolo.desii@powersoft.it, claudio.lastrucci@powersoft.it

³ Audio Labs Systems, Casoli, Chieti, Italy <u>mdicola@lisasystem.com</u>

ABSTRACT

A closed feedback loop transducer system dedicated to very low frequency reproduction can be used in several different applications. The use of a feedback control loop can be very helpful to overcome some of the well known transducer limitations and to improve some of the acoustical performances of most of subwoofer system.

The feedback control of this system is based on a differential pressure control sensor. The entire system control is performed by a "Zero Latency DSP" application, specifically designed for this purpose in order to be able to process the system with real time performances. Practical applications to real world examples are being shown with design details and some test results.

1. INTRODUCTION

Low frequency reproduction systems that use Electrodynamic transducers are based upon very well established approach from a long time. Substantial improvements in design, materials, manufacturing process have been achieved, but, despite large increase in performances, some weak points, due to the nature of the principle of operation still remain unaddressed.

Aging of the compliance of the suspension system, nonlinearities, trade-off between efficiency. performances and applicability in acoustic designs are a limiting factor for several low frequency loudspeaker designs. In addition to this, limitations in repeatability in the manufacturing process is another weak point, hard to overcome even in the most accurate production process. A very desirable low frequency transducer designed for outstanding performances could usually require electromechanical parameters that sometime could fight against the real feasibility. Moreover, this transducer could have, sometime, a limited usability in commercial designs since it could be not easy driven by commonly available amplifiers.

Modern technologies, both in amplifiers and in transducers engineering could expand the possibilities of low frequency system design. Generally speaking, the acoustic loading for a low frequency transducer is calculated and optimized upon the transducer Thiele and Small parameters in order to achieve a desired SPL acoustic response together with the achievable linearity performances and the intended bandwidth, and these performances should be easily obtained when connected to any generic power amplifier that acts like a voltage source.

Anyway, besides the ease of use of loudspeaker transducers when applied in the general approach there are also several limitations to accept that cannot be overcome.

We can resume some of these limitations as follows:

- Consistency between different samples
- Consistency during the lifetime

- Reliability when driven close to maximum limit level
- Suitability for any desired acoustic design
- Transduction efficiency.

A very good solution to overcome these limitations and improve the overall performances of low frequency loudspeaker systems could be the use of closed feedback loop methods to control the loudspeaker. This allows to correct and remove as much as possible the unpredictable and undesired effects of "real world " products, and paying attention to maximize performances and in increasing the system efficiency could also compensate for the cost of added complexity in the system design.

In the wide field of "controlled" transducer there can be found a lot of 'prior art'. Several different realization have been seen since the early 70's [1][2][3][4][5][6] [7][8][9][10][11].

Among the various methods to realize feedback loop control, some of them are based on control of cone acceleration. This method has been able to obtain good performances in terms of linearity in closed cabinet designs, but for high SPL applications it does not show very good performances. Several commercial designs rely on such approach however, delivering consistent performances in terms of frequency response, transient response and overall linearity.

An alternative approach to realize feedback controlled loudspeaker could be based on the control of the cone speed. Methods based on speed control are also able to track a reference pre-processed signal that links the speed to desired output transfer function, still not allowing a complete freedom of applicability for acoustic loads different from closed box. Several other methods move around similar approach with limitations of applicability for acoustic loads that differ from closed cabinet.

We can say that one of the best methods presented in the past that features real substantial improvements in linearity is the *ACE Bass* system. This approach includes, in the closed loop, part of the electromechanical parameters of the physical transducer, being able to alter electrically some of those parameters thus being able to adapt the speaker behavior, within some limitations, to the specific acoustic design.

The ACE system is able to do that with some limitations:

- Inability to reduce the virtual moving mass of the synthesized system to values lower than the real one
- Inability to increase the compliance of the synthesized system to values higher than the real one.
- Being based on a "primary side" current feedback signal, the method can be highly affected by the nonlinearities of the physical transducer itself.

Therefore, the system above, even though it allows for several improvements in sound quality, does not leave a full degree of freedom in terms of synthesis options and, at the same time, does not fully account to the acoustic boundaries conditions that involves the transducer operation.

The novel approach proposed here bases its improved degree of performance on the realization of the feedback loop with Differential Pressure Control method.

2. THE NOVEL APPROACH

The presented system consist in a combination of an high-power, high performances Switching Mode Amplifier in conjunction with an embedded DSP that performs the double operation of both managing the loudspeaker system processing and taking care of the Differential Pressure Feedback Loop Control implemented on it. The described system can be matched to any high performances low frequency, high excursion transducer, even though its performances can be maximized if used in conjunction with some specific transducer that have been especially designed for it and that will be shown ahead in the paper. The system is referred as IPAL (Integrated Powered Adaptive Loudspeaker) [12].

One of the most important innovations of this novel approach is the possibility to use it as a configurable system that may be setup and applied to a wide variety of applications. The system can fit several different kinds of subwoofer loudspeaker and generic low frequency reproduction units of any kind. The system is packed-up in a sort of turnkey solution that once installed in a box, can be connected to a computer via a dedicated communication network and then configured, tuned and optimized for the specific application. Anyway, the configuration process can access to the manipulating of several different systems' parameters, but the approach to the system setup may be resumed in two different philosophies that are represented by two different operating modes.

2.1. Virtual Loudspeaker Emulation

The principle of operation of Virtual Loudspeaker Emulation is based on a double loop control structure. The inner loop takes care to built a very predictable and stable electromechanical structure that is able to make it work like a simple Voltage Input / Speed Output model, where the only visible transfer function from outside is the B*l (Force Factor) of the physical motor assembly.

This very stable building block is then able to track a reference model that could be set at the user's wish and that would be the desired Virtual Transducer.

The Virtual Transducer parameters will be based on a numerical model and it is, of course, not prone to any modification over the lifetime.

The behavior of such defined structure is unfortunately unable to react properly to the acoustic boundary conditions, and the Input Voltage - Output Velocity relationship that would take place would be the one of the loudspeaker operating in the "empty space". In other words, the speaker behavior would be completely unaffected by the forces that an acoustic load acts on the radiating surfaces of the transducer.

The Differential Pressure Measurement, senses the difference of pressure between the front and the rear side of the radiating diaphragm due to the acoustic load, and uses this information to alter the behavior of the reference model, therefore according to the real boundary conditions.

This method allows the modeled unit to behave as a real transducer that keeps the desired properties of the numerical design. An additional description could be explained in the following figures.



Figure 1 - Simplified Model of low frequency Electrodynamic Transducer



Figure 2 - Simplified model of a low frequency transducer with the Compliance, Mass and Losses characteristics moved to the primary side of the "*Bl to* 1" transformer.

As it can be seen in figure 1, a typical simplified model of a low frequency electro-dynamic transducer can be easily described in a schematic where all the physical domains can be involved: the electrical domain together with the mechanical and the acoustical domain. The most meaningful values taken into account are Voltage and Current in the electrical domain, Force and Speed in the mechanical domain, Pressure and Volume Velocity in the Acoustical Domain. As shown in figure 2, is possible to bring from the secondary side of the Bl:1 transformer to the primary the most important mechanical parameters: Compliance, Mass and Losses transposing them into their electrical equivalent parameters Lces, Cmes, Res modeled respectively by an Inductance, a Capacitance and a Resistance. If a series of negative resistance and a negative inductance respectively equal to -Re and -Le (figure 3) will be added at the generator output, it is possible to remove from the primary side of the Bl:1 transformer all the elements of the parallel resonant circuit formed by Lces, Cmes and Res and the synthesis of the circuit could be represented in figure 4 where parallel resonant circuit has been eliminated and the active elements with negative values added in series can be easily realized with the contribution of a negative impedance amplifier whose output impedance behavior can be modeled according to the desired -Re and -Le values.





The whole primary side of the Bl:1 transformer has been reduced to a very desirable block where the input signal is feed to a simple buffer amplifier that drives the Bl:1 transformer with a theoretical unlimited current capability. This arrangement basically allows for the diaphragm speed and its relative Volume Velocity to be very predictable and fully independent from the forces that act on the surface of the diaphragm itself, and link the diaphragm movement directly to the input signal.

The demand for pressure, force, and current, imposed by the acoustical boundary conditions in which the transducer diaphragm operates are therefore accomplished by current capabilities of the Voltage Buffer Negative Impedance Amplifier at the primary side of the *Bl*:1 transformer.

Very simply, a voltage input signal corresponds to a speed of diaphragm at the output, and the relative proportionality is given by the l/Bl, were Bl is the force factor of the real transducer.



Figure 4 - Simplified model where at the primary has been eliminated the resonant parallel circuit and a Negative Impedance amplifier realize the negative -Reand -Le output impedance characteristic.



Figure 5 – Simplified model with a set of desired virtual parameters included before the Negative Impedance Voltage Buffer.

As shown in the above figure 5, it's possible now to define a set of "virtual" electrical parameters of a desired transducer to be emulated, here defined as *Rex*, *Lex*, *Cmesx*, *Lcesx*, *Resx*, *Blx*, *and Sdx*.

Giving the Voltage Buffer Negative Output Impedance amplifier a constant gain structure as (Bl/Blx)(Sdx/Sd), it's easy to demonstrate that this equivalent structure will behave acoustically as a transducer with electrical parameters equal to *Rex*, *Lex*, *Cmesx*, *Lcesx*, *Resx*, *Blx* and *Sdx*, where *Sdx* and *Blx* are the Diaphragm Area and the Force Factor of the Virtual Transducer.

Therefore, as already mentioned, the only limitation of this synthesized loudspeaker model is that it does not react to the boundary condition (acoustical pressure) that is acting on the diaphragm surfaces of the real transducer, and in fact, such a model arrangement will behave identically, in terms of Volume Velocity/Input Voltage relationship, in any acoustic loading conditions: empty space, free space, open air or cabinet acoustical load.

Sensitivity of the newly arranged system will be similar to the sensitivity of the desired Virtual Transducer that is being synthesized, despite the actual motor strength and the actual radiating surface featured by the real transducer.



Figure 6 – The complete system model including the feedback loop with the Differential Pressure signal

It's necessary, at this point, to rebuild a relationship between the forces that actually act on the diaphragm surfaces and the input reference model.

A direct measurement of the difference of acoustical pressure acting on the front and rear diaphragm surfaces, weighted by K, the differential pressure sensor gain and by the ratio of the virtual force factor and virtual diaphragm surface is fed back to the emulated transducer model.

A controlled current source proportional to the above value is here able to rebuild the backward effect of the boundary conditions on the reference model.

It must be clear, once again, that:

- *Re, Le, Cmes, Lces, Res, Bl* and *Sd* parameters are referring to the real transducer,
- *Rex, Lex, Cmesx, Lcesx, Resx, Blx* and *Sdx* are parameters referring the Virtual transducer to be emulated.

u and U are Speed and Volume Velocity referring the real transducer and ux and Ux are Speed and Volume Velocity referring the Virtual transducer to be emulated.

Moreover Zar represents the acoustic radiation impedance and K is the differential pressure sensor gain.

2.2. Global Pressure Control

The novel feedback loop controlled loudspeaker system also allows for an alternative control method. This method involves user's substantially different approach if compared with the already described Virtual Transducer Modeling.

This alternative approach still relies on the differential Pressure Control and permits a very large degree of freedom in using the system from user point of view.

Given a system structure as described in figure 7, it's possible to use it to make it operate in order to match a specific target pressure response and use the output differential pressure measured at the surfaces of the moving diaphragm as a control signal. This value resembles the overall acoustical response with exception of the pressure/SPL transfer function that correlates these entities. Being the SPL output of any acoustic signal derived from the relationship between pressure and acoustical impedance, it's straightforward to use a direct measurement of such a pressure signal to make the system track a reference target pressure model.

A closed loop structure that comprises amplifier device, transducer and acoustical influence is therefore realized with a good stability. Benefits in terms of predictability of the system response and reduction of sensitivity from disturbing effects could be substantially appreciated. This approach allows for a consistent method to make the physical system to perform accordingly a defined Target Pressure Model Response set by the user.



Figure 7 - Block Diagram representing the novel system configured for Target Pressure Model tracking

3. SYSTEM APPLICATION

One of the most important innovations of this novel approach is the possibility to use it as a configurable system that may be setup and applied to a wide variety of applications. The system can fit several different kinds of subwoofer loudspeaker and generic low frequency reproduction units of any kind. The system is packed-up in a sort of turnkey solution that once installed in a box, can be connected to a computer via a dedicated communication network and then configured, tuned and optimized for the specific application. Anyway, the user during the configuration process can access to manipulate several different systems' parameters. The approach to the system setup may be resumed in two different philosophies that are represented by two different operating modes.

3.1. Virtual Speaker Model operation Mode

As described in the previous section, one of the possible ways of using the described system is the so called Virtual Speaker Parameter mode. This way of using this feedback controlled Loudspeaker/Amplifier system is the practical implementation of a way to operate that allows the user to configure the *IPAL* system transducer for a specific desired set of parameters. As already mentioned, previous attempts to built practical realizations of feedback loop controlled loudspeakers were also designed to do that. The novel approach here basically consists in having represented the inner speaker model with a numerical model inside the DSP. A block diagram of the complete system is represented in figure 8.



Figure 8 - Block diagram of the complete system while used in Virtual Loudspeaker Mode

The inner DSP model of the speaker can be, in fact, numerically modified in order to make the speaker appear to the user to behave at the user wish. Moreover, the differential pressure feedback control allows for the modification of a wide set of parameters, and since the speaker model is a numerical model, this suggest to realize the user interaction via a computer interface that allows him to modify the loudspeaker model and adjust the speaker parameters. This also permits to use the transducer parameters like additional "Tuning Parameters" while optimizing the system setup. Setting the transducer parameters in order to get the desired acoustic response from the loudspeaker system is a completely different action from applying some specific equalization, because the parameters are implemented involving the differential pressure feedback loop and then subjected to the loop tracking.

🏟 Virtual Speak	er Parameters	Survey of Street, or other			×
Thiele-Sma	II params				
Qes	0,1882	Qms	6,6134	Fs (Hz)	28,17
Qts	0,1830	Vas (m3)	0,2619		
l		Sd (m2)	0,168	Re(ohm)	0,7400
Elettro mec	hanical paran	15			
Cms (m/N)	0,00006607	Mms (Kg)	0,4836	Rms (kg/s)	12,9362
BL (N/A)	18,3410				
		Sd (m2)	0,1680	Re (ohm)	0,7400
Cmes(Farad)	0,00143761	Lces(Henry)	0,02222544	Res(ohm)	26,00394869
				Re (ohm)	0,7400
Loudspeake	er efficiency				
Sound speed (m/s)	45 Ro me	dium density	(Kg/m3)	1,18
Ref SPL (dB)	5	6,853635495		Ref eff. no%	2,9879
On			Ok	Apply	Cancel

Figure 9 - Software interface where the user can set the desired target parameters



Figure 10 – Software control panel interface while operating in Virtual Speaker Model mode



Figure 11 - Effect of increasing VAS of the Virtual Loudspeaker Model from the real value of 261 Lt. to 500, 800, 1000 Lt.

Some simple examples are shown in figures 11 and 12 to visualize the effect of modifying the values of two simple parameters on the Virtual Speaker Model. Figure 11 shows the effect of modifying the Loudspeker *VAS* from the value of 261 Lt. (Actual *VAS* of the physical speaker) to 500, 800 and 1000 Lt. Figure 12 shows an example of the effect of modification of the Virtual Speaker Bl value from 18.3 (actual physical value) to 15, 12 and 9 Tm.



Figure 12 - Effect of decreasing *Bl* of the Virtual Loudspeaker Model from 18.3 (real value) to 15, 12, 9.

3.2. Pressure Model operation Mode

As already explained, pressure model operation is an operation mode in which the *IPAL* system uses the Differential Pressure Signal and a system control variable. The system control panel changes its appearance as can be seen in figure 13 and a specific interface that is shown in figure 14 allows the user to set directly the influence of the control variable on the system operation and set, ultimately, the amount of feedback control that should be used in the system.



Figure 13 - Software control panel interface while operating in Pressure Model mode



Figure 14 - Software interface section where differential pressure control loop parameters can be accessed

As it can be seen from the above picture of the dedicated control panel, the user is able to set the amount of Feedback Loop Signal from a minimum level of -100dB (feedback off) to 0dB and over, also setting the bandwidth limit above which the control signal should start to be reduced from the nominal value. The amount of feedback loop that may be acceptable depends from the specific application and the specific design where the IPAL system is being set. The upper limit of the amount of feedback will be determined by the system Phase Margin Limits and it must be carefully set in order to let the system operate in stable and reliable mode under any condition. From the Pressure Model control panel it is also possible to set the amplifier output impedance. This feature also permits the system to adapt the loudspeaker transducer to any specific acoustical design. The output impedance could have a bandwidth of operation as well. The optimal setting of these parameters must be still carefully set under the respect of phase margin requirements in order to get the necessary stability of operation.

Once the system has reached the desired level of control with the appropriate and safe stability conditions, the final target response that the low frequency system should have could be reached with appropriate equalization and filtering that may be applied in the input equalizer via a dedicated user interface as showed in figure 15.



Figure 15 - Input DSP software Eq Panel interface

In the following pictures there are reported some interesting test results that have been gathered from a 18" transducer optimized for *IPAL* mounted in a common direct radiating bass reflex loudspeaker box having a volume of 120 Lt and tuned at 36 Hz. There are some 4 cycles tone burst measurements that have been placed at some relevant frequency points: 25Hz, on the first resonant peak on the loudspeaker system impedance, 36 Hz where was the tuning frequency, 50Hz where approximately the second resonant peak was located and then 80Hz and 120Hz, two frequencies of the upper bass range, one located well inside the system bandwidth while the second represents the upper usable frequency limit.

The following diagrams report the test results captured with digital oscilloscope reporting at the same time the input signal on one track and the output pressure measured in front of the cone surface in the other. Not all of them are using the same vertical scale, and this can easily understood and taken into consideration.

From the following measurements it can be observed the effect of the use of the Differential Pressure Feedback control loop, used in Pressure Model mode. The benefit in terms of control of the cone movement is evident and appreciable to the upper frequency limit.

It must be noted that the behavior of this subwoofer without the control loop activated represents a general behavior of such kind of loudspeaker systems, very common to the most of bass reflex subwoofers. Time response behavior of many subwoofer systems is, in fact, quite similar to the system used in the example.



Figure 16 - 25Hz, 4 cycles, sine burst input signal with NO Pressure Control Loop



Figure 18 - 36Hz, 4 cycles, sine burst input signal with NO Pressure Control Loop



Figure 17 - 25Hz, 4 cycles, sine burst input signal <u>with</u> Pressure Control Loop



Figure 19 - 36Hz, 4 cycles, sine burst input signal <u>with</u> Pressure Control Loop



Figure 20 - 50Hz, 4 cycles, sine burst input signal with NO Pressure Control Loop



Figure 22 - 80Hz, 4 cycles, sine burst input signal with NO Pressure Control Loop



Figure 21 - 50Hz, 4 cycles, sine burst input signal <u>with</u> Pressure Control Loop



Figure 23 - 80Hz, 4 cycles, sine burst input signal <u>with</u> Pressure Control Loop



Figure 24 - 120Hz, 4 cycles, sine burst input signal with NO Pressure Control Loop



Figure 25 - 120Hz, 4 cycles, sine burst input signal <u>with</u> Pressure Control Loop

3.3. Example of transducers designed for the Differential Pressure Feedback controlled system

It could be of some interest, for those interested, to mention some details about sample loudspeaker transducers that have been designed and optimized to maximize the system performances. Particularly, these sample loudspeaker transducers are manufactured by B&C Speakers.

Even though this system can be matched to many different kind of loudspeaker, it is very important to point out, anyway, some of the relevant features that a transducer should have in order to maximize the benefit of being coupled to this novel feedback controlled systems.

At first it could to be noted, in fact, that one of the most important advantages of using a switching amplifier stage, usually referred as "Class D" amplifier, is the ability of this kind of power amplifier to manage any kind of Voltage to Current relation at its output, up to the full quadrature between them, practically managing a fully reactive load without any problem. Moreover, where the load is very reactive, the power that is bounced back from the loudspeaker to the amplifier output, in Class D operation, it could be easily recovered and recycled into the power supply rails. This kind of situation would be not possible to be managed by a traditional analog output stage instead, because the power dissipation would be extremely high in that case.

The logic consequence for this is that the loudspeaker to match to the system must be a speaker whose characteristics could be maximized in terms of overall efficiency making it features a very high "Motor Strength" and then an inherently high efficiency loudspeaker that will exhibit a very much reactive electrical behavior. If this last characteristic would be maximized, the amplifier would mostly manage reactive power that is bounced back and forward between the amplifier and the loudspeaker not creating substantial real power dissipation, increasing the overall power efficiency.

Moreover, a very high motor strength over a log excursion capability it would be essential to keep efficient control on the voice coil and cone movement. Some very good transducer to mach this feedback loop controlled loudspeaker would be then shown here. They represent some fine example of use of this system.

The following loudspeaker parameter sets represents very fine example of 18" and 21" loudspeaker that are specifically optimized for this system and allows for maximizing the overall performances.



Figure 26 - IPAL 18 Sample Loudspeaker

18" Sample Loudspeaker Parameters:

Fs	47 Hz
Re	1.37 Ohm
Sd	1210 cm2
Vas	84 Lt
Qms	6.3
Qes	0.23
Bl	22.15 Tm
Mms	283g



Figure 27 - IPAL 21 Sample Loudspeaker

21" Sample Loudspeaker Parameters:

Fs	28.2 Hz
Re	0.74 Ohm
Sd	1680 cm2
Vas	261 Lt
Qms	6.59
Qes	0.19
Bl	18.34 Tm
Mms	484g

3.4. System design considerations

Being the system heavily relying on the inner loop effectiveness, a very predictable and "flat" Bl vs. displacement characteristic of the used transducer is very desirable. Thus a specifically designed motor structure should be a major requirement, even though not mandatory.

Moreover, taking into account that a proper active cancellation is necessary, and being the Re and Le in the inner loop assumed as constants, is therefore necessary that these two parameters would be very stable in the used transducer.

Although some correction of Re vs. temperature is viable, these parameters should be to be varying as less as possible in order to avoid the necessity of direct temperature measurement of the voice coil or the tracking of it with a sophisticated thermal model.

Both the above constrains are achievable in a transducer that has been optimized for very high efficiency, having a $(Bl)^2/Re$ value that is above the usual ranges.

Differential Pressure sensing is also a critical point, since the effect of the acoustical load is derived by a "single point" measurement. At low frequencies, were wavelength of reproduced signals are sufficiently larger than the physical dimensions of the transducer itself, and most of the dimensions of the acoustic system in which the transducer is installed it is relatively easy to keep the distance from the location of the sensor to the centre of emission of the primary transducer relatively small (respect to the shortest operating frequency wavelength). In this case, negligible errors are introduced by the pressure control loop. However, upper frequencies where of the physical transducer behavior is influenced by directivity or by other high order phenomena, are not to be considered as range of operability of the system, or, at last, it could not be required that the system behavior will strictly follow the synthesized model.

It's to be considered anyway that the whole system relies on a "lumped parameters circuit", and consequently its behavior is subjected to some obvious limitations of such assumption. It has to be said that under normal operating conditions like the installation in a relatively compact subwoofer system or into a low frequency unit, the model assumptions are usually widely satisfied.



Figure 27 - IPAL Amplifier Module with DSP inside



Figure 28 – Differential Pressure Sensor

DSP Settings Parameter	ers	Ph	Physical Speaker Parameters			
Z Current limiter	Thiele-Small params					
Time costants (ms)	Threshold(W) Time costants (ms)	Qes	0,1388	Qts	0,1367	
24 Attack \ 10000	700 Attack \ 10000	Qms	9,1251			
		Vas(m3)	0,2617	Sd (m2)	0,35	
Current clamp	Brownout	Fs (Hz)	28,87	Re(ohm)	0,3	
hreshold(A) Time costants (us)	Threshold(V) Time costants (us)					
100 Release 120	Electro mechanical params					
		Cms (m/N)	0.00001521	Rms (kg/s)	39,7373	
Excursion limiter	Time costants (ms)	BL (N/A)	27,9952	Sd (m2)	0,3500	
Inreshold(mm) [20 Gain (Unmich	n) 4 [Release 25]	Mms (Kg)	2,0000	Re (ohm)	0,3000	
Threshold(C ¹) 70 KGain 1000 Sate parameters Hold time(ms) 100 Threshold(dl	0 Over current protection Threshold(A) 120 3) 70 Attenuation (dB) 12	Calculated Cmes(Farad LCes(Henry)	0,00255190	odel Res(ohm) Re (ohm)	19,72281013	
nergy save Threshold(V) 50	Activation time(s) 30	Loudspeak Sound speed	ker efficiency d (m/s)	Ro medium	density (Kg/m3	
dvanced parameters		345		[1,18		
Virt. Speaker RP(mOhm) 2300 S	iafe mode Threshold(W) 20	Ref SPL (dB)	98,492252601	Ref eff. no%	4,3574	
		1			_	
FW VER: 1.3.0, IPAL MOD 1.0	,0		Confirm	Cancel	O.Safe m	

Figure 28 – A view of the inner system setting parameters interface where also physical transducer parameters are set

4. REFERENCES

- Karl Erik Stahl "Synthesis of Loudspeaker Mechanical Parameters by Electrical Means: A new Method for Controlling Low-Frequency Loudspeaker Behavior" – AES Journal, Vol. 29, September 1981
- [2] David S. Hall "Design Considerations for an Accelerometer-Based Dynamic Loudspeaker Motional Feedback System" – Presented at the 87th Convention, New York, USA, October 1989
- [3] Stanislaw Drozdowski "Sound Reproducing Systems with Hall Effect Motional Feedback" – U.S. Patent 4,821,328, Apr., 11, 1989
- [4] William Miller "Loudspeaker with Motional Feedback" U.S: Patent 4,609,784, Sep. 2, 1986
- [5] Al-Ali et al. "Loudspeaker System Feedback Control for Improved Bandwidth and Distortion Reduction" – U.S. Patent 6,584,204 B1, Jun. 24, 2003
- [6] Harada et al. "Optical Motional Feedback" U.S. Patent 4,207,430, Jun. 10, 1980
- [7] Knud E. Bakgaard "Loudspeaker Motional Feedback" U.S. Patent 4,180,706, Dec. 25, 1979
- [8] Guido Odilon Maurits D'Hoogh "Loudspeaker having a Voice Coil and a Piezoelectric Feedback Transducer" – U.S. Patent 3,941,932, Mar. 2, 1976
- [9] Tosshiyuki Goto "Feedback Amplifier Distortion-Cancelling Circuit" – U.S. Patent 3,889,060, Jun. 10, 1975
- [10] Ivan Bekey "Sound Reproduction System" U.S. Patent 3,009,991, Nov. 21, 1961
- [11] Pasi Veli Matias Nuutinmaki "Capacitive Motional Feedback for Loudspeakers", <u>www.servospeaker.com</u>
- [12] Claudio Lastrucci "System for Acoustic Diffusion" – U.S. Patent App. U.S. 2010/0172516